



Frequency transformations in multiple dimensions

DS-GA 1013 / MATH-GA 2824 Mathematical Tools for Data Science

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Calculus (complex numbers)

Linear algebra (orthogonality, basis, projections)

Fourier series

The sampling theorem

Discrete Fourier transform

Square-integrable functions defined on a hyperrectangle $\mathcal{I} := [a_1, b_1] \times \ldots \times [a_d, b_d] \subset \mathbb{R}^d$

Inner product:

$$\langle x, y \rangle := \int_{\mathcal{I}} x(t) \overline{y(t)} dt.$$

Goal: Extension of frequency representations to multidimensional signals

Multidimensional sinusoid

$$a\cos\left(2\pi\langle f,t
ight.
angle+ heta
ight)$$

The frequency and time indices are now *d*-dimensional

Multidimensional sinusoid

If we move in any direction v orthogonal to f

$$a\cos\left(2\pi\langle f,t+v
ight.
angle+ heta
ight)=a\cos\left(2\pi\langle f,t
ight.
angle+ heta
ight)$$

In the direction of f the function is periodic with period $1/||f||_2$

$$\begin{aligned} \operatorname{a}\cos\left(2\pi\left\langle f,t+\frac{1}{||f||_{2}}\frac{f}{||f||_{2}}\right\rangle +\theta\right) &=\operatorname{a}\cos\left(2\pi\left\langle f,t\right\rangle +2\pi\frac{\left\langle f,f\right\rangle}{||f||_{2}^{2}}+\theta\right) \\ &=\operatorname{a}\cos\left(2\pi\left\langle f,t\right\rangle +2\pi+\theta\right) \\ &=\operatorname{a}\cos\left(2\pi\left\langle f,t\right\rangle +\theta\right) \end{aligned}$$

 $\cos\left(2\pi\left\langle \begin{bmatrix} 0\\5\end{bmatrix},\begin{bmatrix} t_1\\t_2\end{bmatrix}\right\rangle\right)$



 $\cos\left(2\pi\left\langle \begin{bmatrix} 10\\0 \end{bmatrix}, \begin{bmatrix} t_1\\t_2 \end{bmatrix}\right\rangle\right)$



 $\cos\left(2\pi\left\langle \begin{bmatrix} 3\\4\end{bmatrix},\begin{bmatrix} t_1\\t_2\end{bmatrix}\right\rangle\right)$



 $\cos\left(2\pi\left\langle \begin{bmatrix} 8\\-6\end{bmatrix},\begin{bmatrix} t_1\\t_2\end{bmatrix}\right\rangle\right)$



Multidimensional complex sinusoids

Complex sinusoid with frequency $f \in \mathbb{R}^d$:

$$\exp(i2\pi\langle f,t\rangle) := \cos(2\pi\langle f,t\rangle) + i\sin(2\pi\langle f,t\rangle)$$

We can express any phase using positive / negative frequencies as in 1D

$$\cos(i2\pi\langle f,t\rangle + \theta) = \frac{\exp(i2\pi\langle f,t\rangle + i\theta) + \exp(-i2\pi\langle f,t\rangle - i\theta)}{2}$$
$$= \frac{\exp(i\theta)}{2}\exp(i2\pi\langle f,t\rangle) + \frac{\exp(-i\theta)}{2}\exp(-i2\pi\langle f,t\rangle)$$

Multidimensional complex sinusoids

Can be expressed as product of 1D complex sinusoids

$$\exp(i2\pi\langle f, t \rangle) := \exp\left(i2\pi \sum_{j=1}^{d} f[j]t[j]\right)$$
$$= \prod_{j=1}^{d} \exp(i2\pi f[j]t[j])$$

2D complex sinusoids

From now on d = 2:

$$f = \begin{bmatrix} f_1 \\ f_2 \end{bmatrix}$$
$$t = \begin{bmatrix} t_1 \\ t_2 \end{bmatrix}$$

$$\exp(i2\pi\langle f,t\rangle) := \exp(i2\pi(f_1t_1 + f_2t_2))$$
$$= \exp(i2\pi f_1t_1)\exp(i2\pi f_2t_2)$$

2D complex sinusoids

Family of complex sinusoids on $[0, T) \times [0, T)$

$$\phi^{ ext{2D}}_{k_1,k_2}\left(t_1,t_2
ight) := \exp\left(rac{i2\pi k_1 t_1}{T}
ight) \exp\left(rac{i2\pi k_2 t_2}{T}
ight), \qquad k_1,k_2 \in \mathbb{Z}$$





 $\phi^{2D}_{0.5} + \phi^{2D}_{0,-5}$ 2







 $\scriptstyle \phi^{\rm 2D}_{10,0} + \phi^{\rm 2D}_{-10,0}$ 2







$$\frac{\phi^{\rm 2D}_{{\bf 3},4}{+}\phi^{\rm 2D}_{-{\bf 3},-4}}{2}$$







 $\phi^{\rm 2D}_{\rm 8,-6}{+}\phi^{\rm 2D}_{\rm -8,6}$ 2



Orthogonality of multidimensional complex sinusoids

The family of complex sinusoids with integer frequencies

$$\phi_{k_1,k_2}^{\text{2D}}\left(t_1,t_2\right) := \exp\left(\frac{i2\pi k_1 t_1}{T}\right) \exp\left(\frac{i2\pi k_2 t_2}{T}\right), \qquad k_1,k_2 \in \mathbb{Z},$$

is an orthogonal set of functions on the interval $[0, T] \times [0, T]$

Proof

We have

$$\phi_{k_{1},k_{2}}^{\text{2D}}(t_{1},t_{2})=\phi_{k_{1}}(t_{1})\phi_{k_{2}}(t_{2}),$$

so that

$$\left\langle \phi_{k_{1},k_{2}}^{2\mathsf{D}},\phi_{j_{1},j_{2}}^{2\mathsf{D}}\right\rangle = \int_{t_{1}=0}^{T}\int_{t_{2}=0}^{T}\phi_{k_{1}}(t_{1})\phi_{k_{2}}(t_{2})\overline{\phi_{j_{1}}(t_{1})\phi_{j_{2}}(t_{2})}\,\mathsf{d}t_{1}\,\mathsf{d}t_{2} = \left\langle \phi_{k_{1}},\phi_{j_{1}}\right\rangle \left\langle \phi_{k_{2}},\phi_{j_{2}}\right\rangle = 0$$

as long as $j_1
eq k_1$ or $j_2
eq k_2$

If
$$j_1 = k_1$$
 and $j_2 = k_2$
 $\left\langle \phi_{k_1,k_2}^{2D}, \phi_{k_1,k_2}^{2D} \right\rangle = \left\langle \phi_{k_1}, \phi_{k_1} \right\rangle \left\langle \phi_{k_2}, \phi_{k_2} \right\rangle$
 $= T^2 \quad \text{so } \left\| \left| \phi_{k_1,k_2}^{2D} \right\| \right\| = T$

2D Fourier series

Fourier series coefficients of a function $x \in \mathcal{L}_2[0, T]$

$$\hat{x}[k_1, k_2] := \left\langle x, \phi_{k_1, k_2}^{\text{2D}} \right\rangle$$
$$= \int_{t_1=0}^{T} \int_{t_2=0}^{T} x(t_1, t_2) \exp\left(-\frac{i2\pi k_1 t_1}{T}\right) \exp\left(-\frac{i2\pi k_2 t_2}{T}\right) \, \mathrm{d}t_1 \, \mathrm{d}t_2$$

The Fourier series of order k_{c1} , k_{c2} is defined as

$$\mathcal{F}_{k_{c1},k_{c2}}\left\{x\right\} := \frac{1}{T^2} \sum_{k_1 = -k_{c1}}^{k_{c1}} \sum_{k_2 = -k_{c2}}^{k_{c2}} \hat{x}[k_1,k_2] \phi_{k_1,k_2}^{\text{2D}}.$$

Non-invasive medical-imaging technique

Measures response of atomic nuclei in biological tissues to high-frequency radio waves when placed in a strong magnetic field

Radio waves adjusted so that each measurement equals 2D Fourier coefficients of proton density of hydrogen atoms in a region of interest

Data



Recovered image



Data



Recovered image



Data



Recovered image



Data



Recovered image



A signal defined on the 2D rectangle $[0, T] \times [0, T]$ is bandlimited with a cut-off frequency k_c if

$$x(t_1, t_2) = \frac{1}{T^2} \sum_{k_1 = -k_c}^{k_c} \sum_{k_2 = -k_c}^{k_c} \hat{x}[k_1, k_2] \exp\left(\frac{i2\pi k_1 t_1}{T}\right) \exp\left(\frac{i2\pi k_2 t_2}{T}\right)$$

Equispaced grid

$$X_{[N]} := \begin{bmatrix} x \left(\frac{0}{N}, \frac{0}{N}\right) & x \left(\frac{0}{N}, \frac{T}{N}\right) & \cdots & x \left(\frac{0}{N}, T - \frac{T}{N}\right) \\ x \left(\frac{T}{N}, \frac{0}{N}\right) & x \left(\frac{T}{N}, \frac{T}{N}\right) & \cdots & x \left(\frac{T}{N}, T - \frac{T}{N}\right) \\ \cdots & \cdots & \cdots \\ x \left(T - \frac{T}{N}, \frac{0}{N}\right) & x \left(T - \frac{T}{N}, \frac{T}{N}\right) & \cdots & x \left(T - \frac{T}{N}, T - \frac{T}{N}\right) \end{bmatrix}$$

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Nyquist-Shannon-Kotelnikov sampling theorem

Any bandlimited signal $x \in \mathcal{L}_2[0, T)^2$, where T > 0, with cut-off frequency k_c can be recovered from N^2 uniformly spaced samples if

 $N \geq 2k_c + 1$,

where $2k_c + 1$ is known as the Nyquist rate

We represent 2D signals as matrices belonging to the vector space of $\mathbb{C}^{N\times N}$ matrices endowed with the standard inner product

$$\langle A,B \rangle := \operatorname{tr}(A^*B), \quad A,B \in \mathbb{C}^{N \times N}$$

Equivalent to dot product between vectorized matrices

Discrete complex sinusoids

The discrete complex sinusoid $\psi_{k_1,k_2}^{2D} \in \mathbb{C}^{N \times N}$ with integer frequencies k_1 and k_2 is defined as

$$\psi_{k_1,k_2}^{2D}[j_1,j_2] := \exp\left(\frac{i2\pi k_1 j_1}{N}\right) \exp\left(\frac{i2\pi k_2 j_2}{N}\right), \qquad 0 \le j_1, j_2 \le N-1,$$

Equivalently

$$\psi_{k_1,k_2}^{\mathsf{2D}} = \psi_{k_1} \psi_{k_2}^\mathsf{T}$$

The discrete complex exponentials $\frac{1}{N}\psi_{k_1,k_2}^{2D}$, $0 \le k_1, k_2 \le N - 1$, form an orthonormal basis of $\mathbb{C}^{N \times N}$

Proof

$$\begin{split} \left\langle \psi_{k_1,k_2}^{\text{2D}},\psi_{l_1,l_2}^{\text{2D}} \right\rangle &= \operatorname{tr}\left(\left(\psi_{l_1,l_2}^{\text{2D}}\right)^* \psi_{k_1,k_2}^{\text{2D}} \right) \\ &= (\psi_{k_1})^* \psi_{l_1} (\psi_{k_2})^* \psi_{l_2} \end{split}$$

Nyquist-Shannon-Kotelnikov sampling theorem

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where $2k_c + 1$ is known as the Nyquist rate

$$\hat{x}[k_1, k_2] = \frac{T}{N^2} \left\langle \psi_{k_1, k_2}^{\text{2D}}, X_{[N]} \right\rangle$$
$$= \frac{T}{N^2} \psi_{k_1}^* X_{[N]} \psi_{k_2}.$$

Proof

For any
$$m_1$$
, m_2 , we have $\phi^{2D}_{m_1,m_2}\left[rac{j_1T}{N},rac{j_2T}{N}
ight] = \psi^{2D}_{m_1,m_2}[j_1,j_2]$, so

$$\begin{aligned} \frac{T}{N^2} \left\langle \psi_{k_1,k_2}^{2\mathsf{D}}, X_{[N]} \right\rangle &= \frac{T}{N^2} \left\langle \psi_{k_1,k_2}^{2\mathsf{D}}, \frac{1}{T} \sum_{m_1=-k_c}^{k_c} \sum_{m_2=-k_c}^{k_c} \hat{x}[m_1, m_2] \psi_{m_1,m_2}^{2\mathsf{D}} \right\rangle \\ &= \sum_{m_1=-k_c}^{k_c} \sum_{m_2=-k_c}^{k_c} \hat{x}[m_1, m_2] \left\langle \frac{1}{N} \psi_{k_1,k_2}^{2\mathsf{D}}, \frac{1}{N} \psi_{m_1,m_2}^{2\mathsf{D}} \right\rangle \\ &= \hat{x}[k_1, k_2] \end{aligned}$$

2D discrete Fourier transform

The discrete Fourier transform (DFT) of a 2D array $X \in \mathbb{C}^{N \times N}$ is

$$\widehat{X}[k_1,k_2] := \left\langle X, \psi_{k_1,k_2}^{2\mathsf{D}} \right\rangle, \qquad \mathsf{0} \le k_1, k_2 \le \mathsf{N}-1,$$

or equivalently

$$\widehat{X} := F_{[N]} X F_{[N]},$$

where $F_{[N]}$ is the 1D DFT matrix

It can be computed efficiently with the FFT (complexity $N^2 \log N$)

Inverse 2D discrete Fourier transform

The inverse DFT of a 2D array $\widehat{Y} \in \mathbb{C}^{N \times N}$ equals

$$Y = \frac{1}{N^2} F^*_{[N]} \widehat{Y} F^*_{[N]}$$

It inverts the 2D DFT

Properties of multidimensional sinusoids

Definition of multidimensional Fourier series

Definition of multidimensional DFT

Sampling theorem in 2D